

Simulation of the SNS DTL BPM High Frequency Performance

R. E. Shafer 3/7/2002

This simulation examines the high frequency signal power of the DTL BPMs as a function of beam energy and beam micropulse shape, for the present design BPM.

===== basic constants =====

$$c := 2.997924 \cdot 10^{10} \text{ cm/sec} \quad Mc2 := 939.28 \text{ MeV}$$

===== beam properties =====

$$Ib := 0.038 \quad f := 402.5 \cdot 10^6 \quad w := 2 \cdot \pi \cdot f$$

$$\text{gam}(E) := 1 + \frac{E}{Mc2} \quad \text{bgam}(E) := \sqrt{\text{gam}(E)^2 - 1} \quad \text{beta}(E) := \frac{\text{bgam}(E)}{\text{gam}(E)}$$

===== BPM =====

$$a := 1.44 \text{ cm, effective radius} \quad L := 3.4 \text{ cm, effective length}$$

$$Z := 50 \text{ ohms} \quad \phi := 60 \text{ degrees, electrode width}$$

$$\text{argBPM}(n, E) := \frac{n \cdot w \cdot L}{2 \cdot c} \cdot \left(1 + \frac{1}{1} \right) \quad \text{argument for BPM with shorted electrodes}$$

===== Bessel factor =====

This is the low-beta effect

$$\text{argBes}(n, E) := \frac{n \cdot w \cdot a}{\text{bgam}(E) \cdot c} \quad \text{Bes}(n, E) := \frac{1}{I_0(\text{argBes}(n, E))} \quad \text{dB change} \\ \text{dBes}(n, E) := 20 \cdot \log(\text{Bes}(n, E))$$

===== Fourier transform of pulse shape =====

$$\text{FWAB(sig)} := 4.472 \cdot 10^{-12} \cdot \text{sig} \quad \text{Full width at base of parabolic micropulse, in terms of the rms width sig(ps).}$$

$$\alpha(n, \text{sig}) := n \cdot \pi \cdot \text{FWAB(sig)} \cdot f$$

$$A_m(n, \text{sig}) := 3 \cdot \left(\frac{\sin(\alpha(n, \text{sig}))}{\alpha(n, \text{sig})^3} - \frac{\cos(\alpha(n, \text{sig}))}{\alpha(n, \text{sig})^2} \right) \quad \text{Fourier harmonic amplitudes for parabolic bunch shape}$$

$$\text{dBpulse}(n, \text{sig}) := 10 \cdot \log(A_m(n, \text{sig})^2) \quad \text{dB change}$$

===== coax cable =====

atten := -2 approx attenuation for 50 m of 1/2" heliax at 400 MHz

dBatten(n) := atten· \sqrt{n} dB change

===== dBm power for basic BPM =====

$V(n, E) := \sqrt{2} \cdot \frac{\text{phi}}{360} \cdot I_b \cdot Z \cdot \sin(\arg \text{BPM}(n, E))$

$pwr(n, E) := 1000 \frac{V(n, E)^2}{Z}$ milliwatts

$\text{dBm}(n, E) := 10 \cdot \log(pwr(n, E))$ dBm output for basic BPM without corrections

===== total power level =====

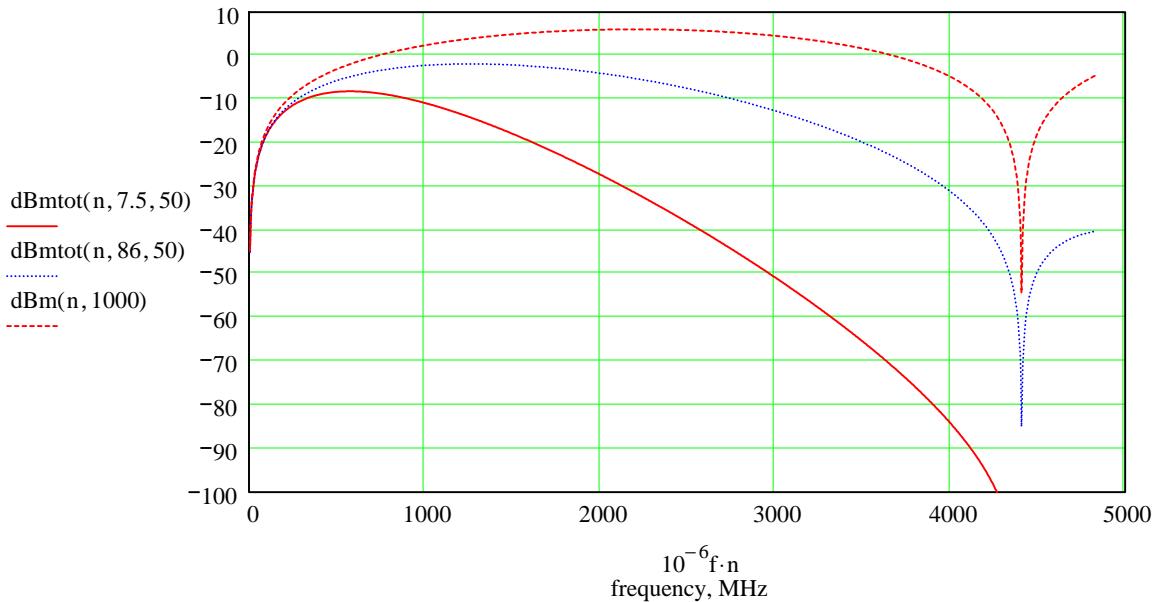
$\text{dBmtot}(n, E, \text{sig}) := \text{dBm}(n, E) + \text{dBBes}(n, E) + \text{dBatten}(n) + \text{dBpulse}(n, \text{sig})$

Example $\text{dBm}(10, 86) = -5.339$ $\text{dBBes}(10, 86) = -12.21$ $\text{dBatten}(10) = -6.325$

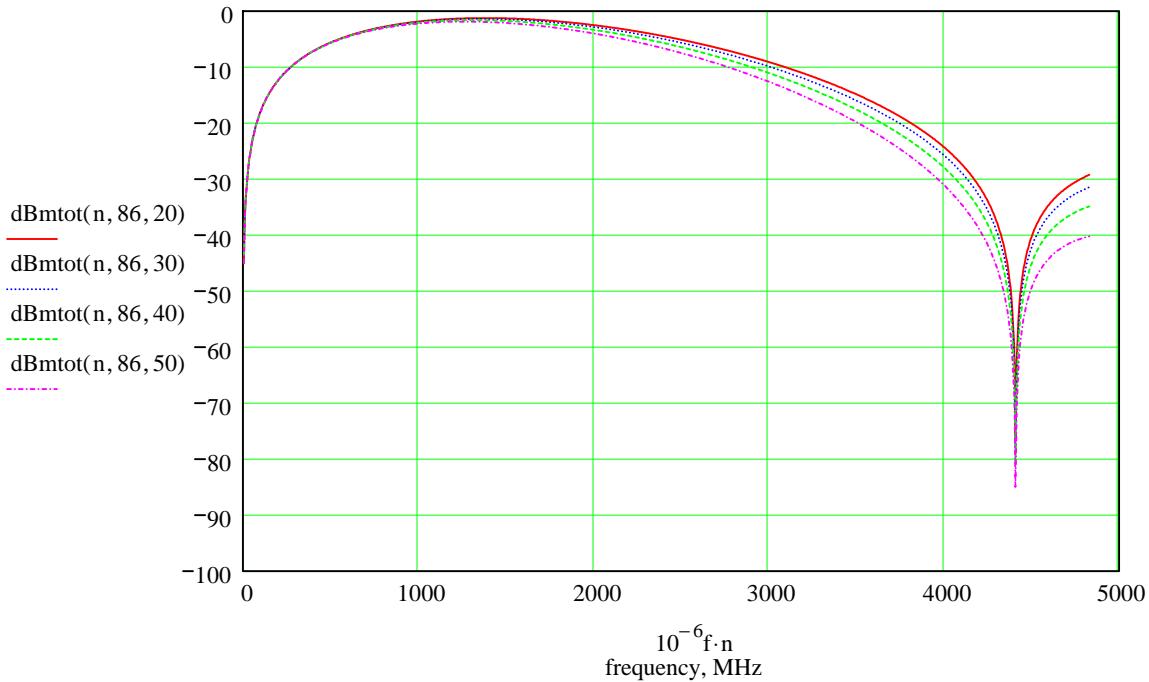
$\text{dBpulse}(10, 50) = -8.004$ $\text{dBmtot}(10, 86, 50) = -31.878$

$n := 0.01, 0.02..12$

Signal power vs. frequency for the basic BPM, and for 7.5 and 86 MeV



Effect of bunch length on output power at 86 MeV



Because the effective length of the BPM is known to only about +/-3%, the signal power above 4 GHz is very uncertain. It is clear that the best high frequency measurements would be between 3 and 4 GHz.

The approximate cutoff frequencies are (using the effective radius of BPM):

$$f_{TE} := \frac{10^{-9} \cdot 3.832 \cdot c}{2 \cdot \pi \cdot a} \quad f_{TE} = 12.697 \quad \text{GHz, TE mode}$$

$$f_{TM} := \frac{10^{-9} \cdot 2.405 \cdot c}{2 \cdot \pi \cdot a} \quad f_{TM} = 7.969 \quad \text{GHz, TM mode}$$

Measurements probably should be limited to about 4 GHz or lower.

Amplitudes of the 10th harmonic (about 4 GHz) are:

$$10 \cdot 10^{-9} \cdot f = 4.025 \quad \text{GHz}$$

$\text{dBmtot}(10, 86, 10) = -24.153$ dBm, 10 ps rms bunch length

$\text{dBmtot}(10, 86, 20) = -25.006$ dBm, 20 ps rms bunch length

$\text{dBmtot}(10, 86, 30) = -26.486$ dBm, 30 ps rms bunch length

$\text{dBmtot}(10, 86, 40) = -28.703$ dBm, 40 ps rms bunch length

$\text{dBmtot}(10, 86, 50) = -31.878$ dBm, 50 ps rms bunch length